

Solid State Amplifiers for Next- Generation Radar Transmitters

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2011 International Symposium on Advanced Radar

Technologies (ISART 2011)

Boulder, Colorado

July 2011

Outline

- Baylor WMCS Program and Activities
- Review of Single-Stage Amplifier Classes
- Advanced Power Amplifier Architectures
- Real-Time Load-Impedance Optimization
- The Way Forward
 - Joint Waveform and Circuit Optimization
 - Wirtinger Calculus for TIPP Systems

Baylor WMCS Program

- Wireless and Microwave Circuits and Systems
- Wireless and Microwave Education and Research in a Caring, Christian Environment
- Launched in 2008.



Research

- Faculty
 - Dr. Charles Baylis, Co-Director, microwave power amplifier design, waveform diversity
 - Dr. Randall Jean, Co-Director, microwave sensors and metrology
 - Dr. Yang Li – antenna design
 - Dr. Robert J. Marks II – computational intelligence
 - Dr. Steve Eisenbarth – wireless networks
 - Dr. Mike Thompson - communications
- Graduate and undergraduate student research and teaching assistants

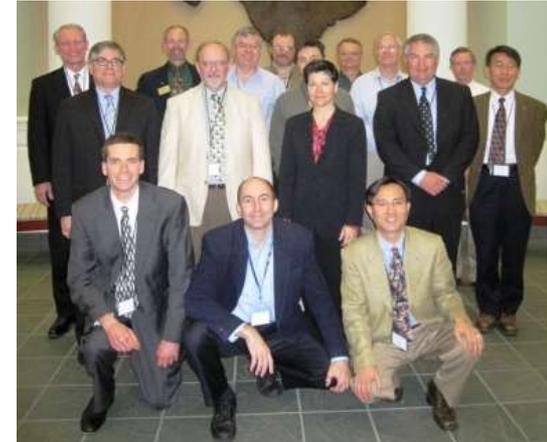
WMCS Teaching Laboratory

- Founded in 2009 with partial sponsorship from Agilent Technologies.
- “Hub” for hands-on teaching activity.
- Provides hands-on component for RF/Microwave Circuits course sequence.



WMCS Advisory Board and Mini-Symposium

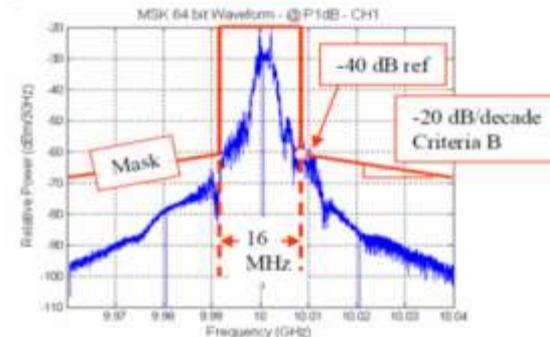
- Industry Advisory Board created in 2009 to assist with educational and research mission.
- Annual Mini-Symposium on Wireless and Microwave Circuits and Systems
 - Student/industry forum
 - 5 universities participated in 2011.



Transmitter Amplifier Constraints

- Amplifiers must transmit large amounts of power with high power efficiency.
- Radar spectrum criteria imposed in the Radar Spectrum Evaluation Criteria (RSEC), which are determined by the National Telecommunications and Information Administration (NTIA).
- Spectral mask outlines the required confines of the signal:

*Reprinted from J. de Graaf, H. Faust, J. Alatishe, and S. Talapatra, "Generation of Spectrally Confined Transmitted Radar Waveforms," Proc. IEEE Conf. on Radar, 2006, pp. 76-83



Sources of Nonlinearity

- A major source of spectral spreading is third-order intermodulation distortion in the amplifier transistor.
- Assume a third-order nonlinear system approximated by

$$v_{out}(t) = a + bv_{in}(t) + cv_{in}^2(t) + dv_{in}^3(t)$$

- Stimulate with a two-tone input signal:

$$v_{in}(t) = A \cos \omega_1 t + B \cos \omega_2 t$$

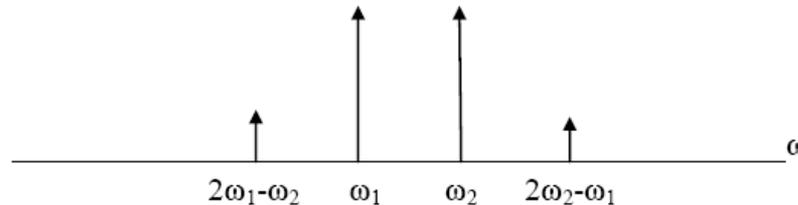
Math Results

$$v_{out}(t) = a + b(A\cos \omega_1 t + B\cos \omega_2 t) + c(A\cos \omega_1 t + B\cos \omega_2 t)^2 + d(A\cos \omega_1 t + B\cos \omega_2 t)^3$$

$$v_{out}(t) = a + b(A\cos \omega_1 t + B\cos \omega_2 t) + c(A^2 \cos^2 \omega_1 t + B^2 \cos^2 \omega_2 t + 2AB \cos \omega_1 t \cos \omega_2 t)^2 + d(A^2 \cos^2 \omega_1 t + B^2 \cos^2 \omega_2 t)(A\cos \omega_1 t + B\cos \omega_2 t)$$

$$v_{out}(t) = a + b(A\cos \omega_1 t + B\cos \omega_2 t) + c(A^2 \cos^2 \omega_1 t + B^2 \cos^2 \omega_2 t + 2AB \cos \omega_1 t \cos \omega_2 t)^2 + d(A^3 \cos^3 \omega_1 t + A^2 B \cos^2 \omega_1 t \cos \omega_2 t + AB^2 \cos \omega_1 t \cos^2 \omega_2 t + B^3 \cos^3 \omega_2 t)$$

Third-Order Intermodulation Terms



Intermodulation Results

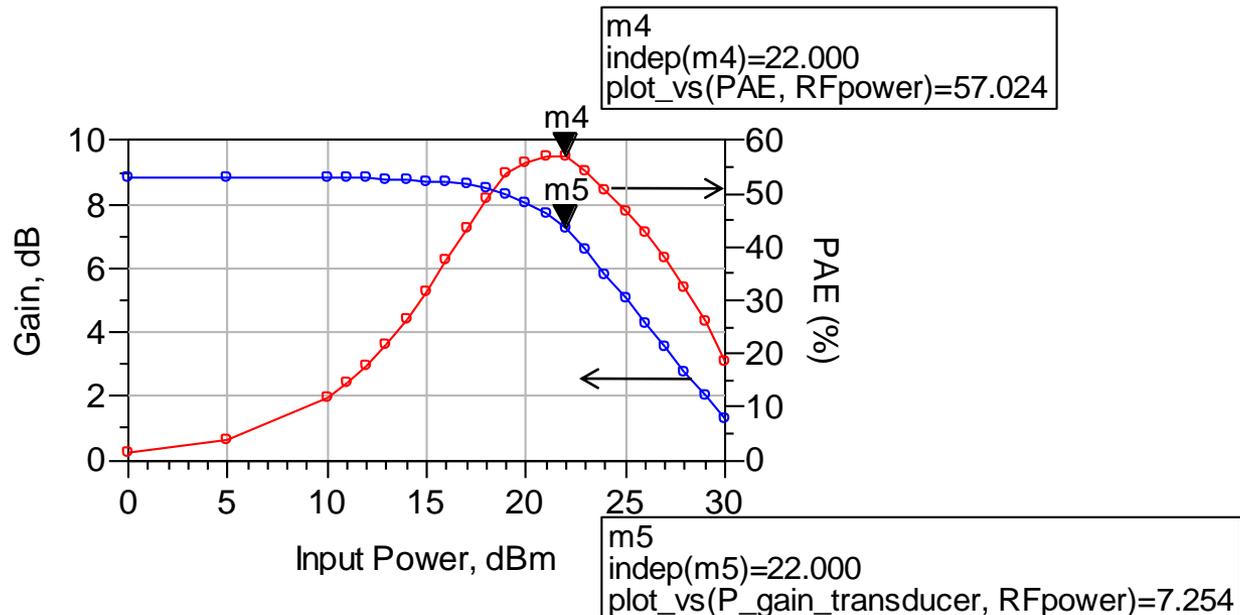
- For a bandpass signal, each frequency at which the signal is nonzero represents a “tone”.
- In general, all pairs of tones intermodulate:
 - In-band distortion
 - Out-of-band distortion (“spectral spreading”)

How to Remove the Sidelobes?

- Filtering?
 - Radar systems often operate in megawatt range.
 - It is difficult to use microstrip cavity filter capabilities over 1 kW.
 - Not cost-effective or practical
- Linearization
 - Remove the sidelobes by making the amplifier more linear.
 - We need to maintain efficiency at the same time.

Linearity vs. Efficiency

- Efficiency increases with output power.
 - GaAs MESFET power amplifier example shown below.
- Linearity decreases with increasing output power for amplitude modulated signals.



Efficiency Measures

- Drain Efficiency:

$$\eta = \frac{P_{out,RF}}{P_{DC}}$$

- Power-Added Efficiency:

$$\eta_{ADD} = \frac{P_{out,RF} - P_{in,RF}}{P_{DC}}$$

- Overall Efficiency:

$$\eta_{OVERALL} = \frac{P_{out,RF}}{P_{DC} + P_{in,RF}}$$

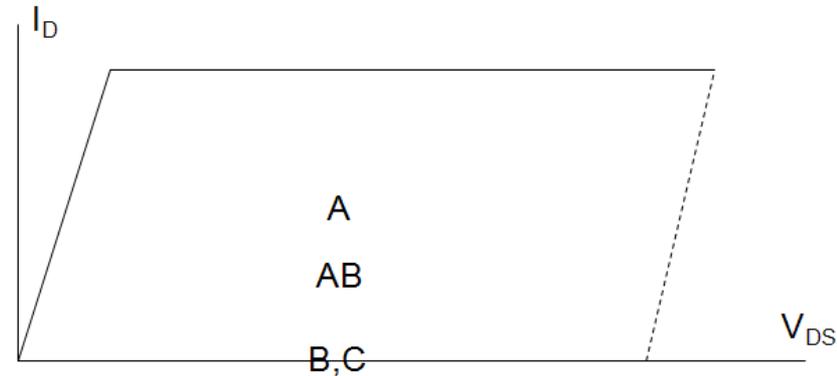
Linearity Measures

- Adjacent Channel Power Ratio (ACPR)
 - Ratio of the power in a specified band outside the signal to the RMS power in the signal*
 - Examines how nonlinearity affects adjacent channels.
- Error Vector Magnitude
 - Vector distance between desired and measured signal vector normalized by the signal amplitude
- Carrier-to-Intermodulation (C/I) Ratio
 - Measured in a two-tone intermodulation test.
 - Raab: C/I should be at least 30 dB for a linear PA.*
- Noise Power Ratio
 - Measures in-band distortion.

*Source: F. Raab, P. Asbeck, S. Cripps, P. Kenington, Z. Popovic, N. Pothecary, J. Sevic, and N. Sokal, "RF and Microwave Power Amplifier and Transmitter Topologies, Part 1," *High Frequency Electronics*, May 2003.

Amplifier Classes

- Class A:
 - Max Drain Efficiency: 50%
 - Best Intrinsic Linearity
- Class B:
 - Max Drain Efficiency: 78.5%
 - Reduction of Linearity
- Class C:
 - Bias below threshold.
 - Higher efficiency but less linearity than B.
- Class E, F: Higher efficiency switching modes.

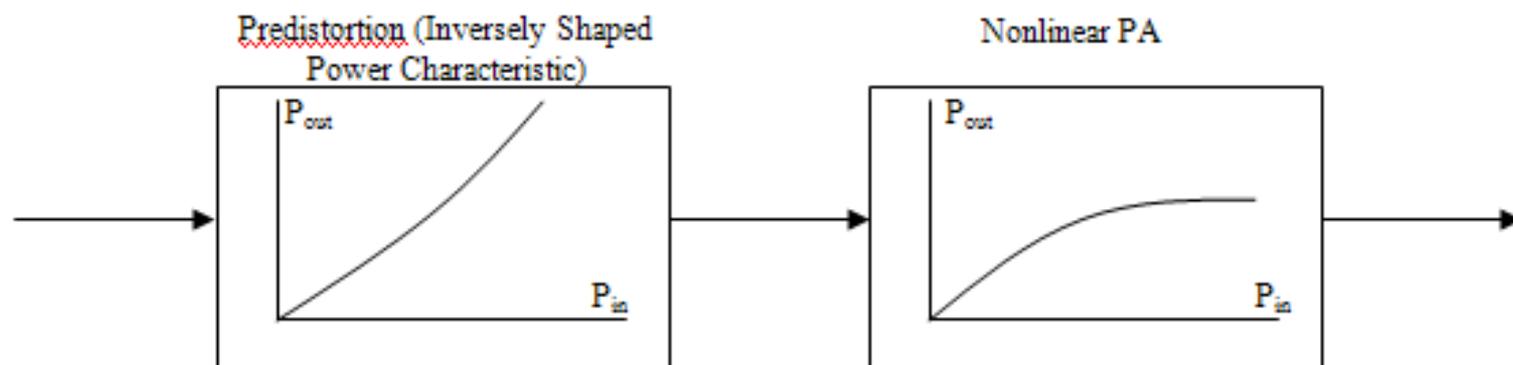


Linearity and Efficiency Configurations

- Acknowledgment: Article in *IEEE Transactions on Microwave Theory and Techniques* and 4-part series of articles in *High Frequency Electronics* by Raab *et al.* provide an excellent survey of different topologies and their advantages.
- Much information from these articles is used in this section.

Predistortion

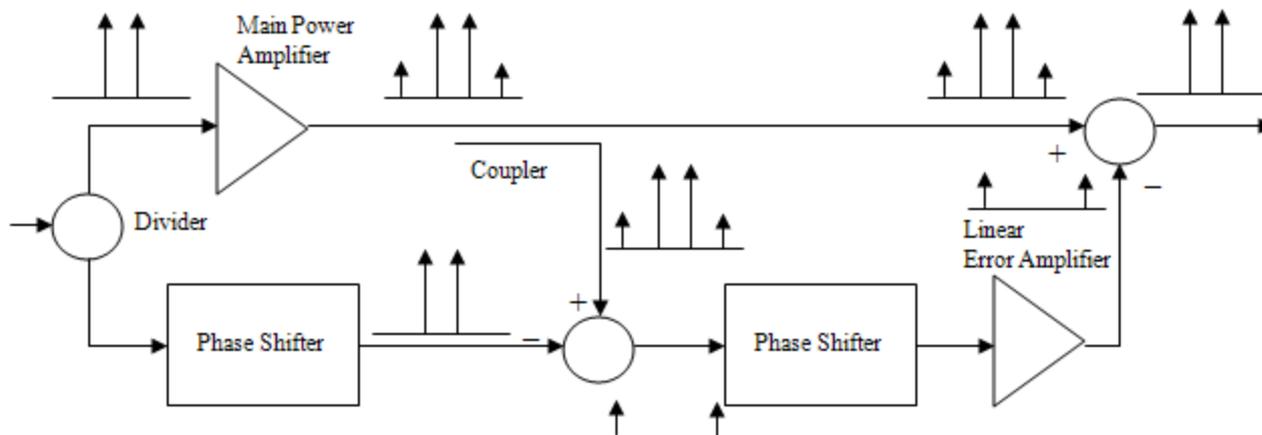
- “Uncompress” the compression by a component with an oppositely shaped compression characteristic.



- Challenging for systems with memory
- Requires adaptive lookup table \rightarrow memory requirements can be large.

Feedforward

- The linear input signal is used as a reference to subtract unwanted spectral components from the output signal.

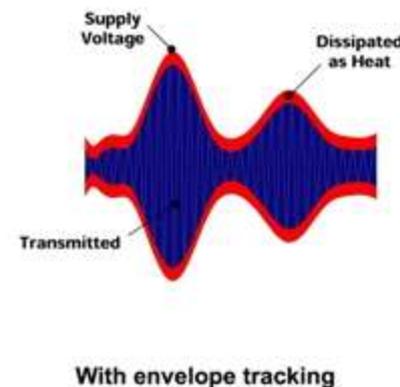
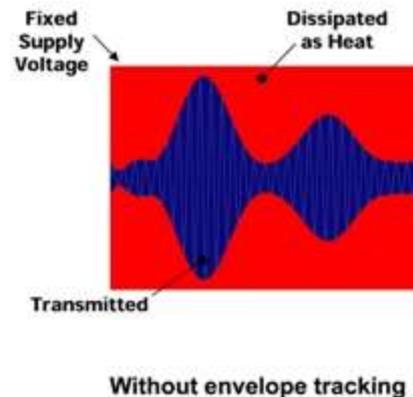
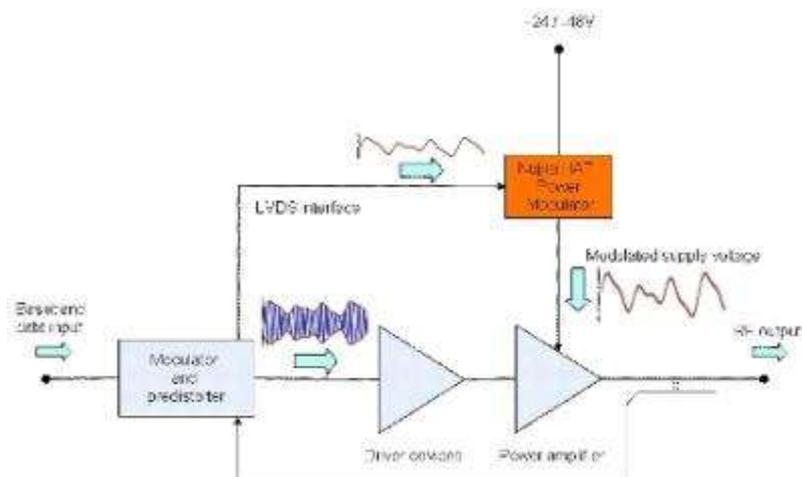


- Linear error amplifier requires additional DC power.
- Combiners also contribute to efficiency decrease.
- Drift is possible; may require a control system.

Acknowledgment: F. Raab, P. Asbeck, S. Cripps, P. Kenington, Z. Popovic, N. Potheary, J. Sevic, and N. Sokal, "RF and Microwave Power Amplifier and Transmitter Topologies, Part 4," *High Frequency Electronics*, November 2003.

Envelope Tracking

- The supply voltage is adjusted based on envelope amplitude.
- The efficiency is improved, but buck/boost converters require additional DC.
- Works well for high peak-to-average-power ratio (PAPR):



*Pictures reprinted from G. Wimpenny, "Improving Multi-Carrier PA Efficiency Using Envelope Tracking," *RF Engineer Network*, <http://www.rfengineer.net>, April 2009.

Additional Source: F. Raab et al., "Power Amplifiers and Transmitters for RF and Microwave"

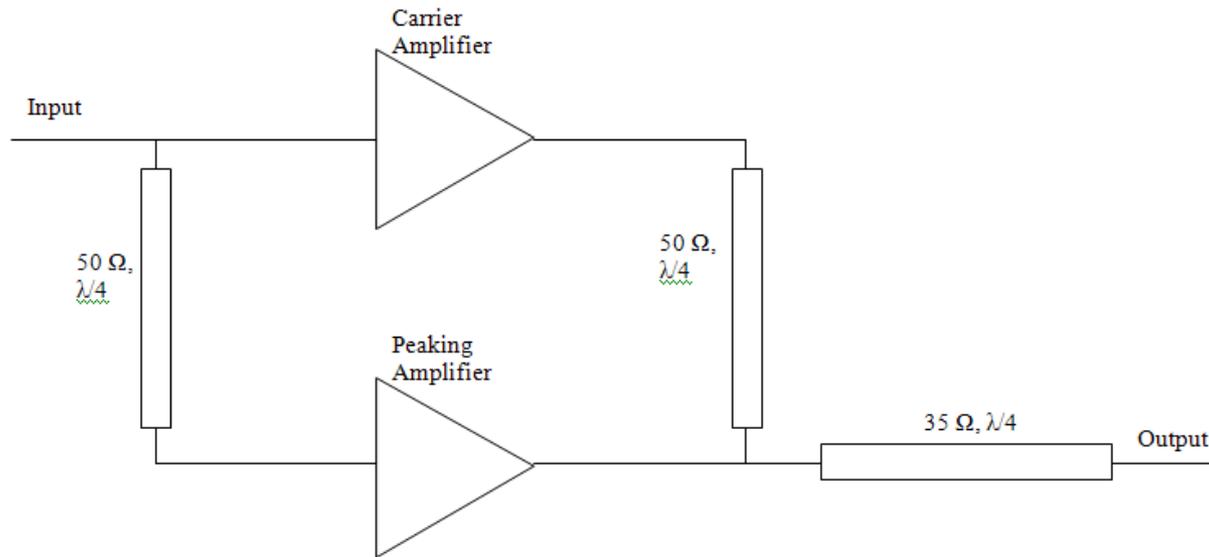
IEEE Transactions on Microwave Theory and Techniques, pp. 814-826, Vol. 50, No. 3, March

Envelope Elimination and Restoration (Kahn Technique)

- The amplitude modulation is removed from the signal and re-inserted after the PA.
- Allows the amplitude to run at optimum efficiency without amplitude distortion.
- Must align amplitude and phase modulation (need low AM-to-PM conversion).*

*Source: F. Raab *et al.*, *IEEE Transactions on Microwave Theory and Techniques*, Vol. 50, No. 3, March 2002.

Doherty



- Carrier Amplifier: Class B
- Peaking Amplifier: Class C
- Peaking amplifier turns on when the signal becomes large.
- Linearity is at Class B level from this design.

LINC (Linear Amplification with Nonlinear Components)

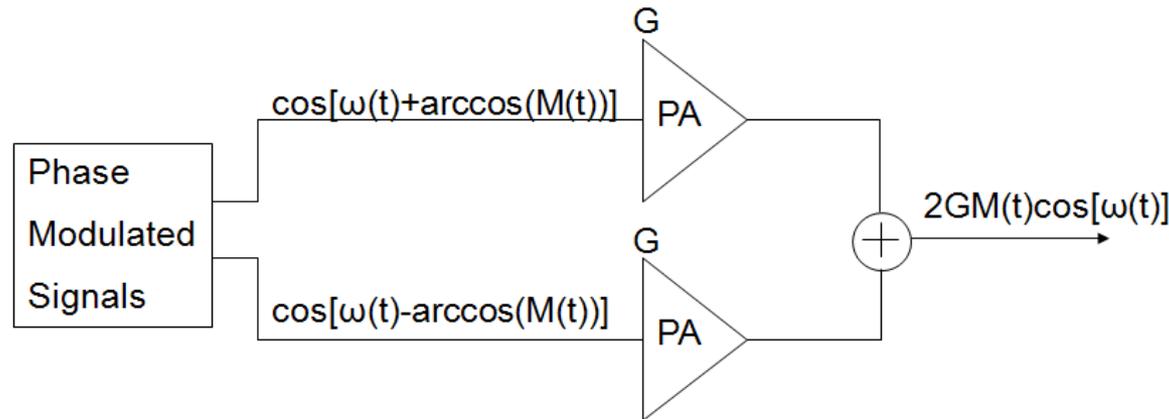
- The amplitude modulation $M(t)$ is “hidden” in the phase and returned to the amplitude after the summer:

$$v_1(t) = \cos(\omega t + \cos^{-1}(M(t)))$$

$$v_2(t) = \cos(\omega t - \cos^{-1}(M(t)))$$

$$v_{out}(t) = G \cos(\omega t + \cos^{-1}(M(t))) + G \cos(\omega t - \cos^{-1}(M(t)))$$

$$v_{out}(t) = 2G \cos \omega t \cos(\cos^{-1}(M(t))) = 2GM(t) \cos \omega t$$



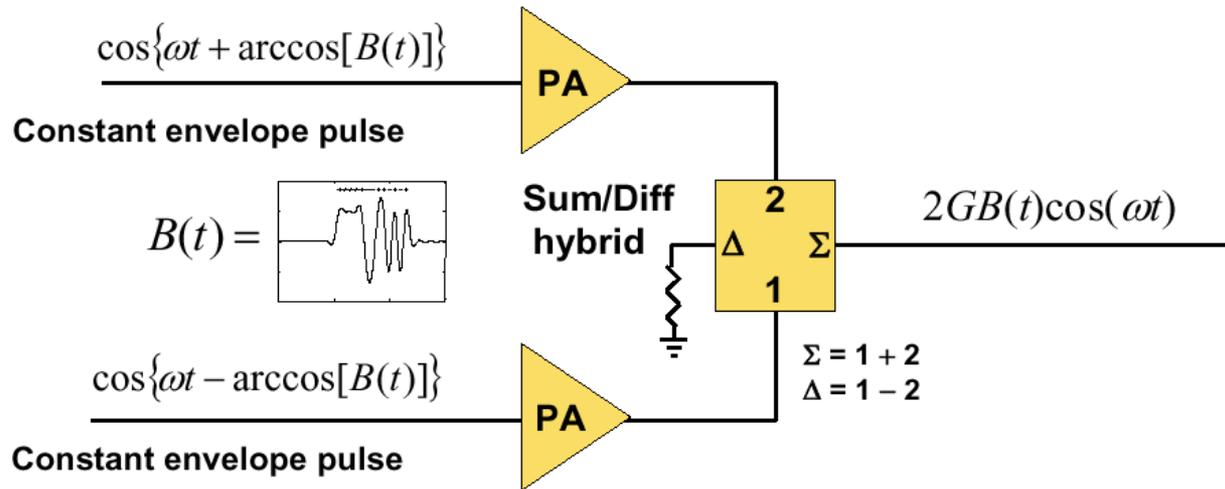
- But how can a summer be implemented?

Implementation Options

- 180-Degree Coupler
- Chireix Outphasing Combiner*
- Linearity and efficiency vary by modulation scheme for each design.
- 180-degree coupler is more robust for linearity.

*H. Chireix, "High Power Outphasing Modulation," *Proceedings of the IRE*, Vol. 23, pp. 1370-1392, November 1935.

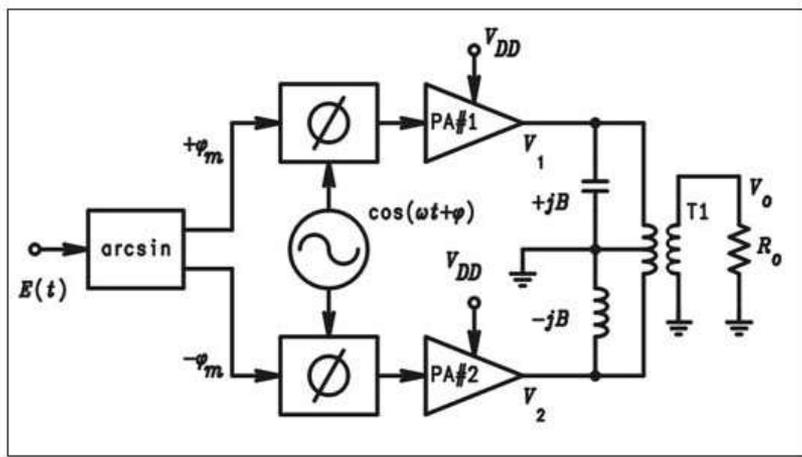
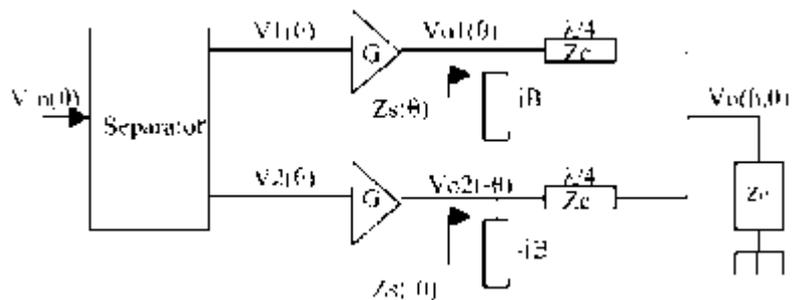
180-Degree Coupler



- It is matched and reciprocal.
- Power can be lost to the fourth-port termination, depending on the modulation scheme.

*Reprinted from J. de Graaf, H. Faust, J. Alatishe, and S. Talapatra, "Generation of Spectrally Confined Transmitted Radar Waveforms," Proc. IEEE Conf. on Radar, 2006, pp. 76-83

Chireix Combiner

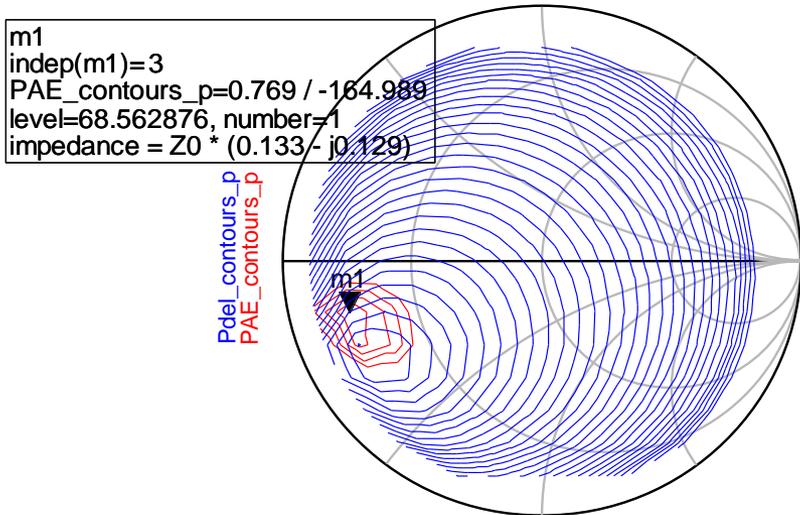


- Combiner ports are not isolated, so the impedances seen by each amplifier stage can change dynamically.

*Reprinted from F. Raab, P. Asbeck, S. Cripps, P. Kenington, Z. Popovic, N. Pothecary, J. Sevic, and N. Sokal, "RF and Microwave Power Amplifier and Transmitter Technologies, Part 1," *High Frequency Electronics*, May 2003.

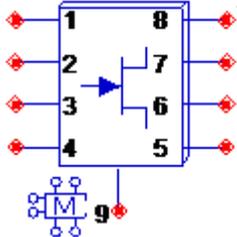
**A. Birafane and A. Kouki, "On the Linearity and Efficiency of Outphasing Microwave Amplifiers," *IEEE Transactions on Microwave Theory and Techniques*, Vol. 52, No. 7, July 2004, pp. 1702-1708.

PHEMT Source/Load Pull

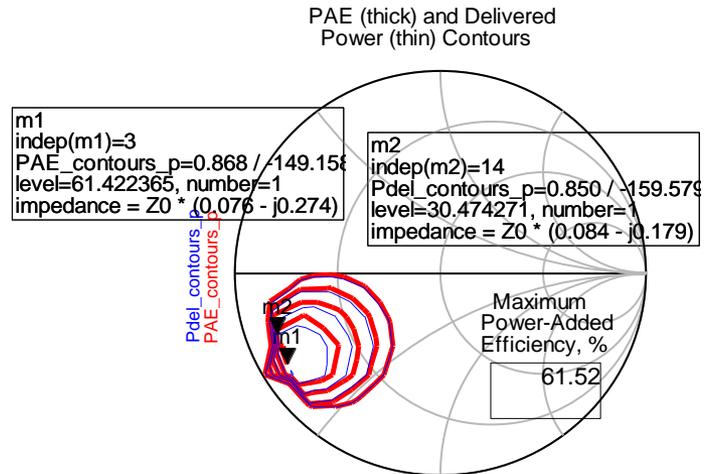


indep(PAE_contours_p) (0.000 to 15.000)
 indep(Pdel_contours_p) (0.000 to 55.000)

Load Pull: Found maximum
 PAE at 68.6% with a source
 impedance of $(6.65 - j6.45)$



Modelithics
 Transistor
 Model

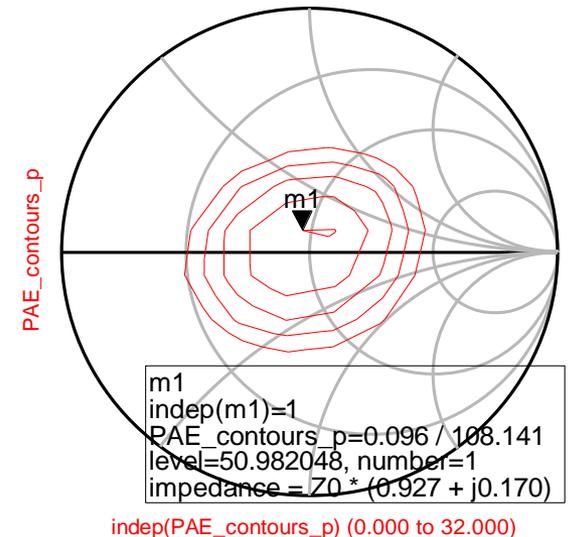
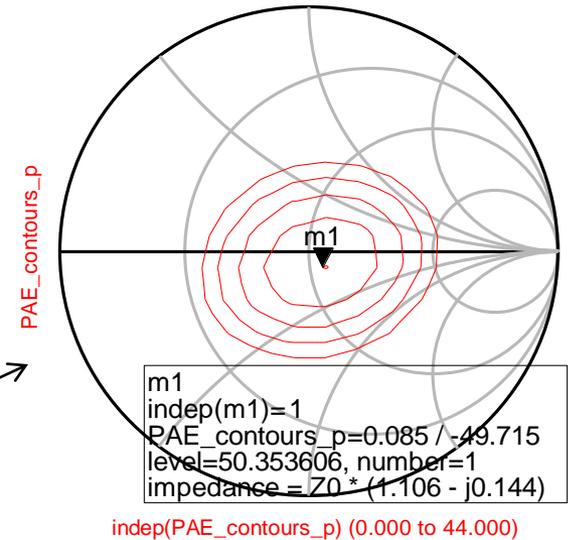


indep(PAE_contours_p) (0.000 to 30.000)
 indep(Pdel_contours_p) (0.000 to 30.000)

Source Pull: Found
 maximum PAE at 61.5%
 with a source impedance
 of $(4.2 - j8.95)$

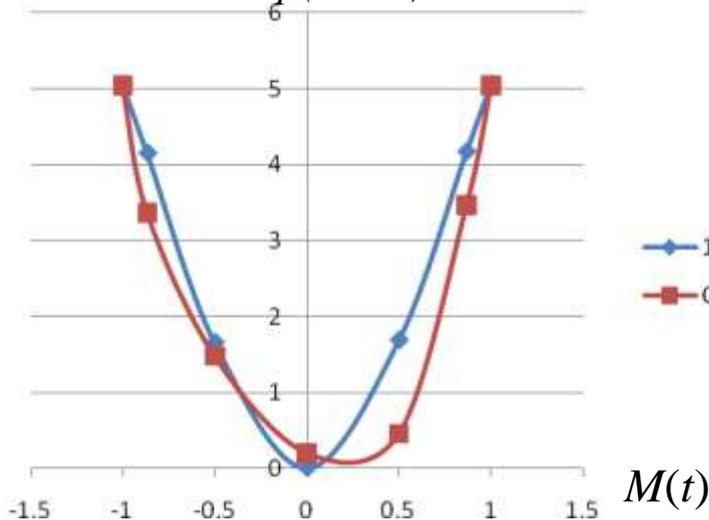
Load-Pull Efficiency Comparison

- Both Designs: Maximum Efficiency near 50 ohms
- Test with CW ($M(t) = 1$)
- Maximum PAE for Chireix design = 50%
- Maximum PAE for 180-degree coupler design = 51%

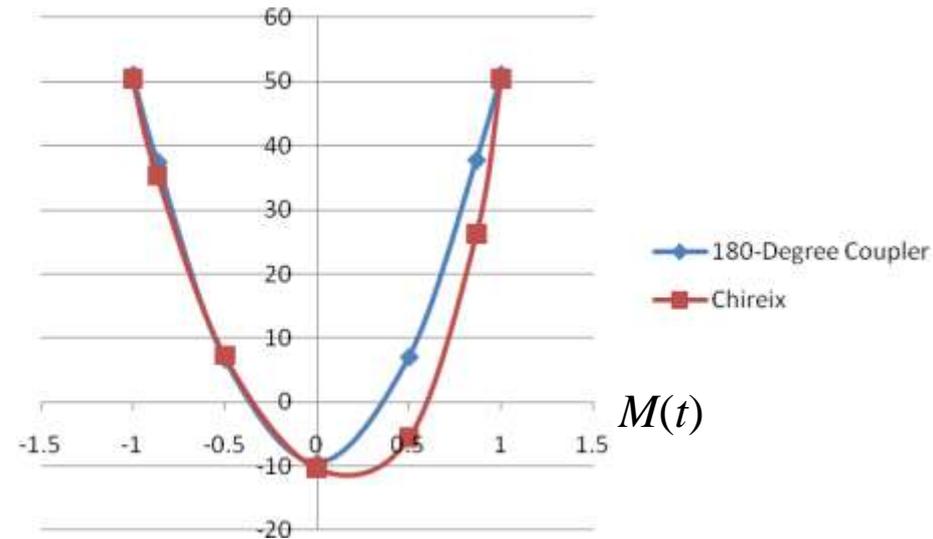


GaAs PHEMT Amplifier for Different $M(t)$ Levels

G_T (W/W)



PAE (%)



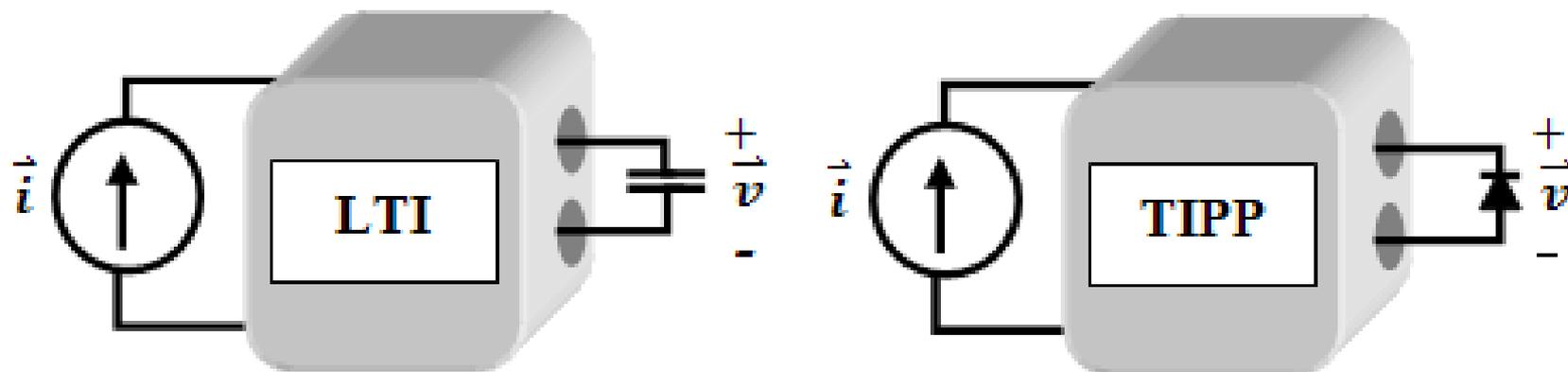
- 180-degree coupler is perfect parabola \rightarrow Excellent linearity
- Chireix demonstrates linearity flaws.

Joint Optimization: The Way Forward

- State-of-the-art approaches to improving spectral conformity have traditionally included separate examination of
 - Circuit design
 - Waveform design
- The technology and theory now exist to *simultaneously* optimize both!
- Knowing the circuit nonlinearities speeds the optimization → Wirtinger Calculus for TIPP Systems.

TIPP Systems

- Assume a time invariant periodicity preservation (TIPP) system.



LTI: All currents and voltages oscillate at the same frequency.

TIPP: All currents and voltages are periodic with the same period (harmonic levels can change).

TIPP Systems

- Assume a TIPP Operator \mathcal{Z} :

$$v(t) = \mathcal{Z} \{i(t)\}$$

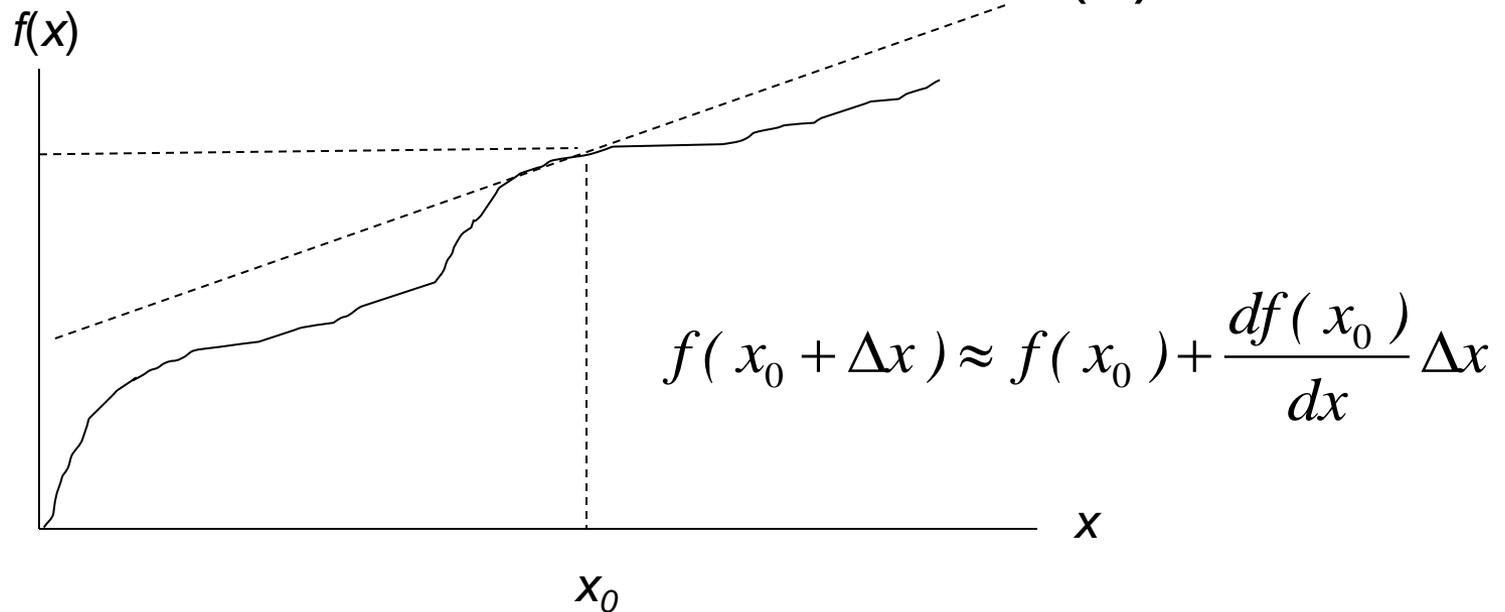
- There is a corresponding operator on the vectors of Fourier coefficients:

$$\vec{v} = \mathbf{Z} \vec{i}$$

- For a particular “operating point” large signal, \mathbf{Z} is a matrix.

Affine Approximation

- Consider a nonlinear function $f(x)$:



- Affine approximation around the operating point of a nonlinear function

Wirtinger Calculus for TIPPP Systems

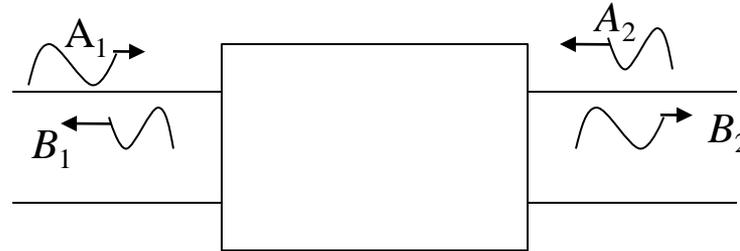
$$v(t) + \Delta v(t) \approx Z\{i(t)\} + \frac{\partial v(t)}{\partial i(t)} \Delta i(t) + \frac{\partial v(t)}{\partial i^*(t)} \Delta i^*(t)$$

- In terms of the Fourier series coefficient vectors:

$$\bar{v} + \Delta \bar{v} \approx [Z] \bar{i} + [J_{\bar{v}}(\bar{i})] \Delta \bar{i} + [J_{*\bar{v}}(\bar{i})] \Delta \bar{i}^*$$

- The TIPPP parameters give an affine approximation around a nonlinear operating point.

Agilent X-Parameters^{1*}



¹X-parameters is a registered trademark of Agilent Technologies.

$$B_{ef} = X_{ef}^{(F)}(|A_{11}|)P^f + \sum_{g,h} X_{ef,gh}^{(S)}(|A_{11}|)P^{f-h} a_{gh} + \sum_{g,h} X_{ef,gh}^{(T)}(|A_{11}|)P^{f+h} a_{gh}^*$$

Each X parameter is a function of $|A_{11}|$.

**C. Baylis *et al.*, "Going Nonlinear," *IEEE Microwave Magazine*, April 2011.

$X^{(S)}$

$P = e^{j\angle A_{11}}$ provides phase correction for harmonic conversion.

ef, gh

Arrival Port

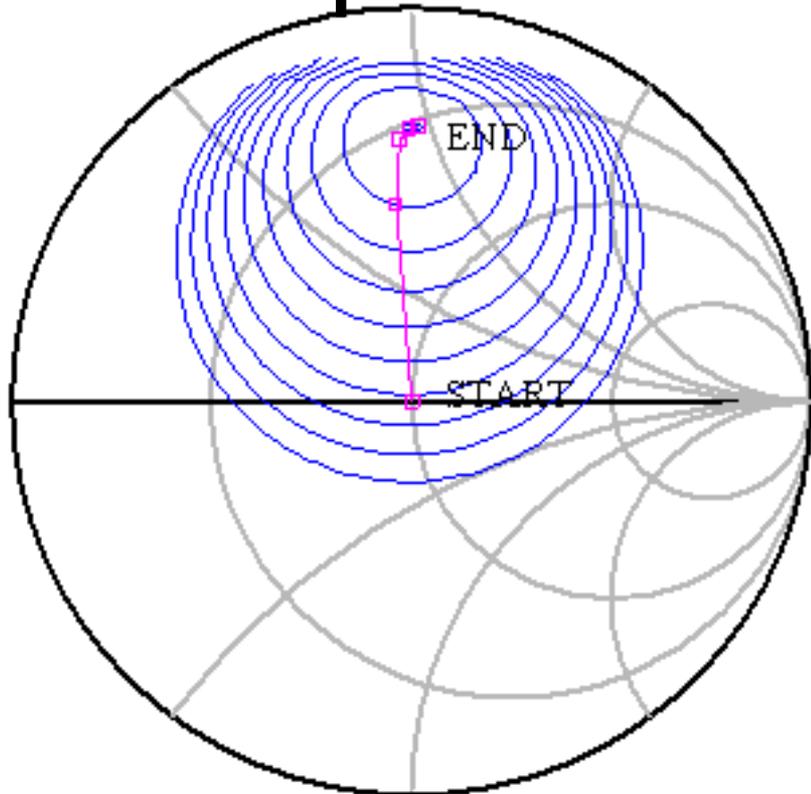
Arrival Harmonic

Departure Port

Departure Harmonic

*D. Root, "A New Paradigm for Measurement, Modeling, and Simulation of Nonlinear Microwave and RF Components," Presentation at Berkeley Wireless Research Center, April 2009

Fast Load-Impedance Optimization Algorithm*



- Traditional:
 - 400 Γ states
 - Maximum Power = 22.76 dBm
- Steepest Ascent:
 - 17 Γ states
 - Maximum Power = 22.72 dBm
- Accurate results for small number of simulations

*C. Baylis, L. Dunleavy, S. Lardizabal, R.J. Marks II, and A. Rodriguez, “Efficient Optimization Using Experimental Queries: A Peak-Search Algorithm for Efficient Load-Pull Measurements,” Accepted for Publication in *Journal of Advanced Computational Intelligence and Intelligent Informatics*, September 2010.

Conclusions

- Spectral spreading from radar systems must be mitigated, but not at the cost of system efficiency.
- Several useful design approaches exist for linearity and efficiency improvement.
- An apparent solution is in joint waveform and circuit optimization with the Wirtinger calculus.
- An approach and test platform for real-time load-pull and waveform optimization is under development at Baylor University.

Acknowledgments

- Dr. Robert J. Marks II, Baylor University
- Baylor Research Assistants: Loria Wang, Josh Martin, Matthew Moldovan, Hunter Miller, Robert Scott
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- Agilent Technologies, for cost-free loan of the Advanced Design System software.
- Maury Microwave for donation of ATS Software DLL Libraries.
- Modelithics, Inc., for donation of model libraries.
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